

GENERAL DESCRIPTION

OB2361 is a high performance flyback adaptive multi-mode PWM controller, optimized to achieve high efficiency in the universal input range and full loading range with effective system cost.

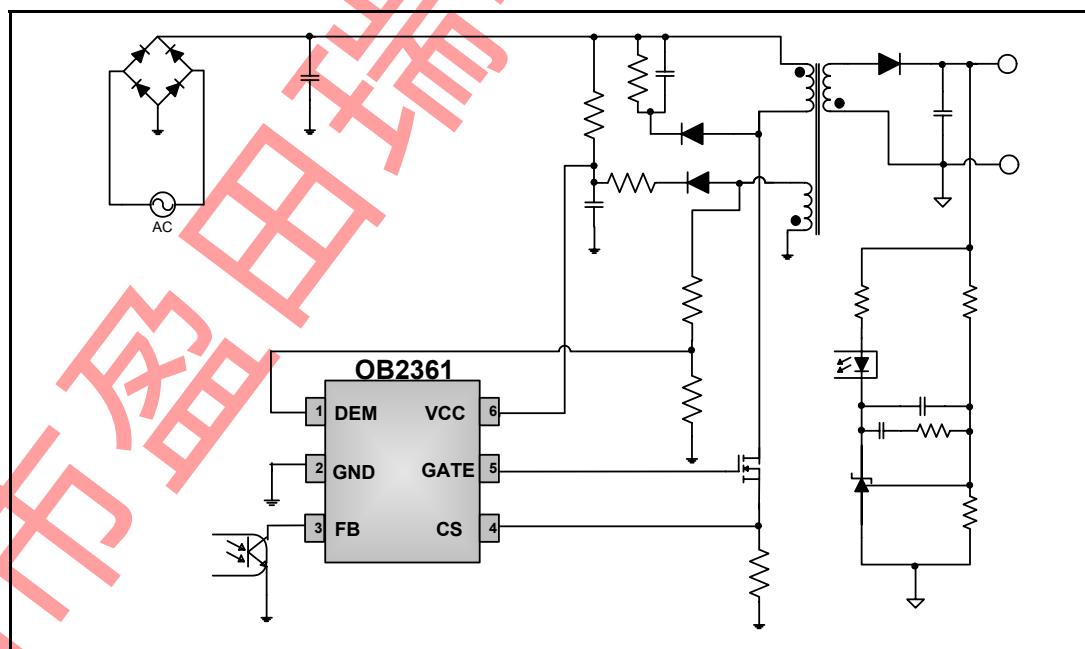
At full loading, the IC operates in fixed frequency (52KHz) CCM mode in the low line input voltage and it operates in QR mode in high line input voltage. In this way, high efficiency in the universal input range at full loading can be achieved.

At normal load condition, it operates in QR mode. To minimize switching loss, the maximum switching frequency in QR mode is internally limited to 95 KHz. When the loading goes low, it operates in PFM mode with valley switching for high power conversion efficiency. When the load is very small, the IC operates in 'Extended Burst Mode' to minimize the standby power loss. As a result, high conversion efficiency can be achieved in the whole loading range.

OB2361 offers comprehensive protection coverage including Cycle-by-Cycle Current Limiting (OCP), VCC Under Voltage Lockout (UVLO), Output Over Voltage Protection (OVP), Over Load Protection (OLP), VCC Over Voltage Protection (VCC OVP), VCC Clamp, and Gate Clamp. OB2361 also features protections against pin open and short conditions on selected pins.

OB2361 is offered in SOT23-6 packages.

TYPICAL APPLICATION



FEATURES

- Multi-Mode Operation:
 - At Full Loading, Fix Frequency (52KHz)
 - CCM Operation @ Low Line and QR Mode Operation @ High Line Input
 - Quasi-Resonant Operation at Normal Load
 - Pulse Frequency Modulation (PFM) Operation at Light Load
 - Burst Mode at No Load
- Adaptive Frequency Shuffling and Slope Compensation @ Fix Frequency CCM Mode
- 4ms Soft-start
- Over Load Protection
- Over Current Protection
- Pin Open/Short Protection
- Output Over Voltage Protection

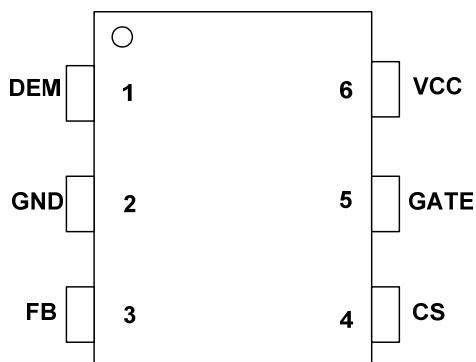
APPLICATIONS

- Offline AC/DC flyback converter for
- Power Adapter and Open-frame SMPS
 - Set-Top Box Power Supplies
 - NB/DVD

GENERAL INFORMATION

Pin Configuration

The pin map of OB2361 in SOT23-6 package is shown as below.



Package Dissipation Rating

Package	R _{θJA} (°C/W)
SOT23-6	200

Absolute Maximum Ratings

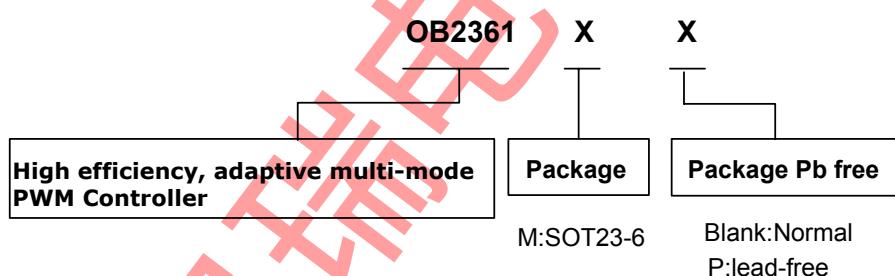
Parameter	Value
VCC	29 V
VCC Clamp Current	10 mA
FB Input Voltage	-0.3 to 7V
CS Input Voltage	-0.3 to 7V
DEM Input Voltage	-0.7 to 7V
Maximum Operating Junction Temperature T _J	150 °C
Min/Max Storage Temperature T _{stg}	-55 to 150 °C
Lead Temperature (Soldering, 10secs)	260 °C

Note: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum-rated conditions for extended periods may affect device reliability.

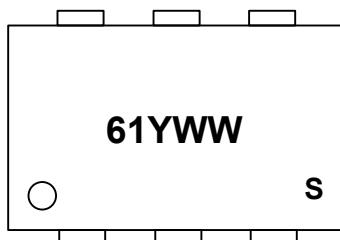
Ordering Information

Part Number	Description
OB2361MP	SOT23-6, Pb-free in T&R

Note: All Devices are offered in Pb-free Package if not otherwise noted.



Marking Information

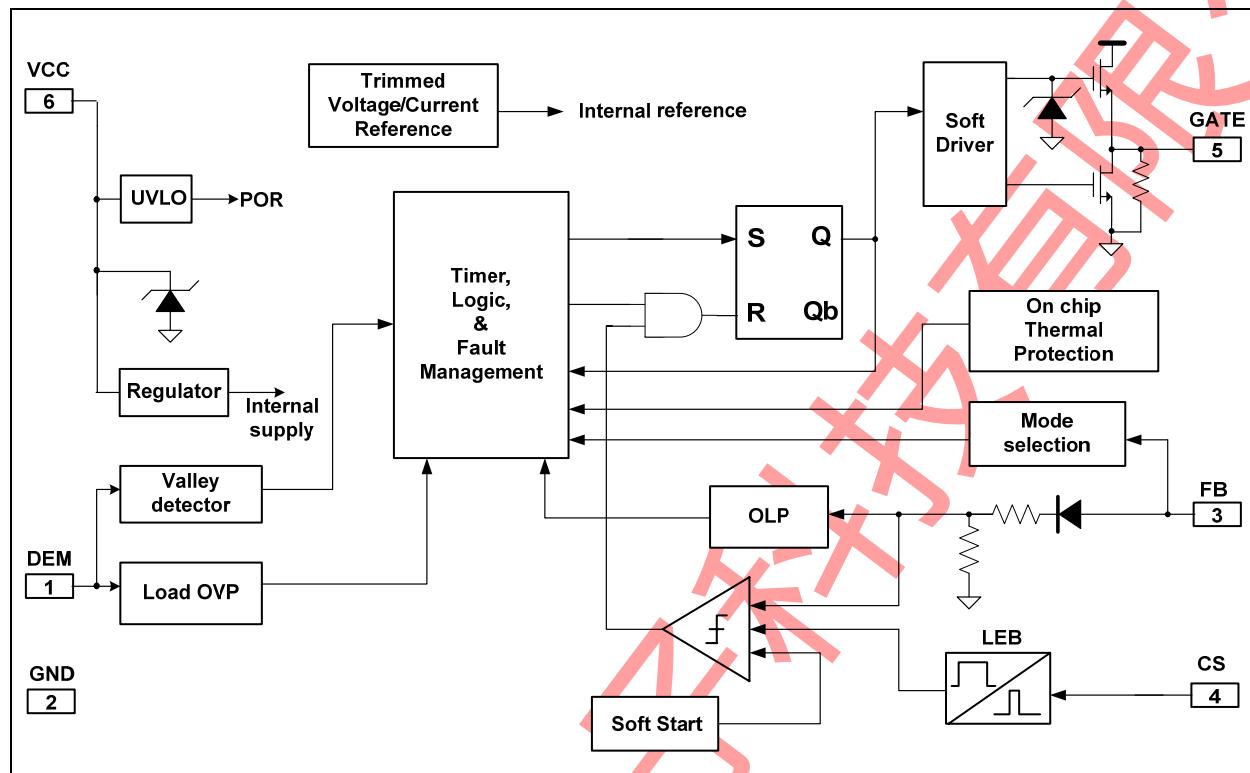


Y:Year Code(0-9)
 WW:Week Code(01-52)
 S:Internal Code(Optional)

TERMINAL ASSIGNMENTS

Pin Num	Pin Name	I/O	Description
1	DEM	I	Transformer core demagnetization detection pin. This pin is also used for output over voltage protection (OVP).
2	GND	P	Ground for internal circuitry.
3	FB	I	Feedback input pin. PWM duty cycle is determined by voltage level into this pin and current-sense signal level at Pin 3.
4	CS	I	Current sense input.
5	GATE	O	Totem-pole gate drive output for power MOSFET.
6	VCC	P	Chip DC power supply pin.

BLOCK DIAGRAM



RECOMMENDED OPERATING CONDITION

Symbol	Parameter	Min	Max	Unit
VCC	VCC Supply Voltage	10	25	V
T _A	Operating Ambient Temperature	-20	85	°C

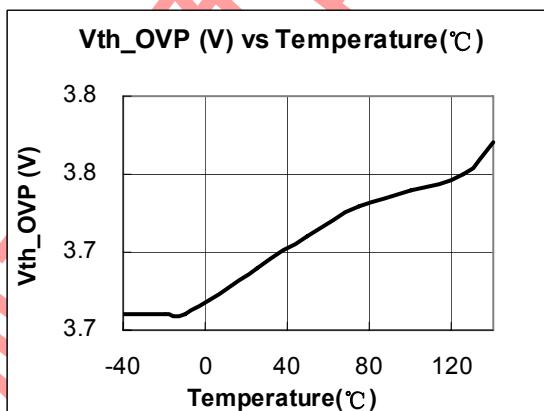
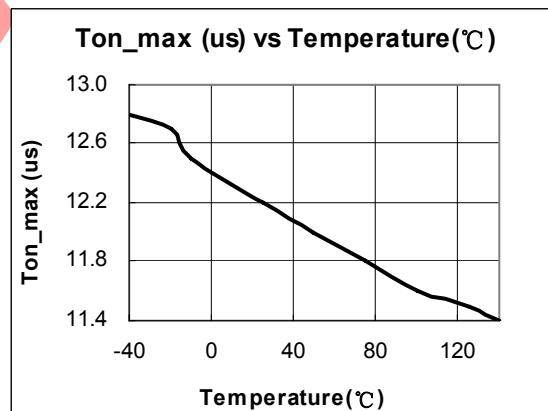
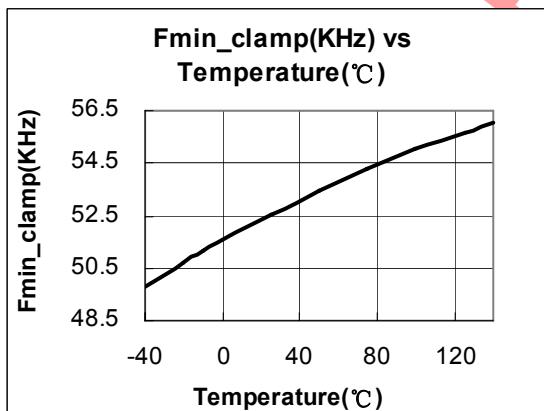
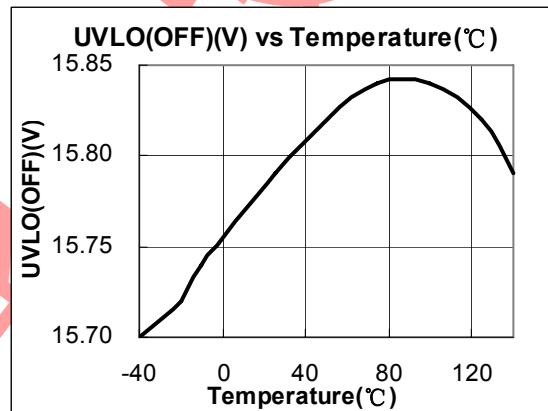
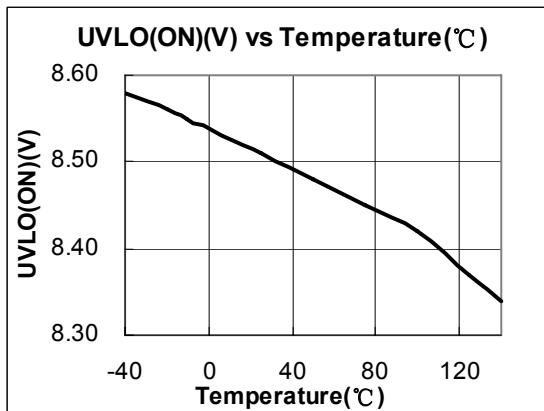
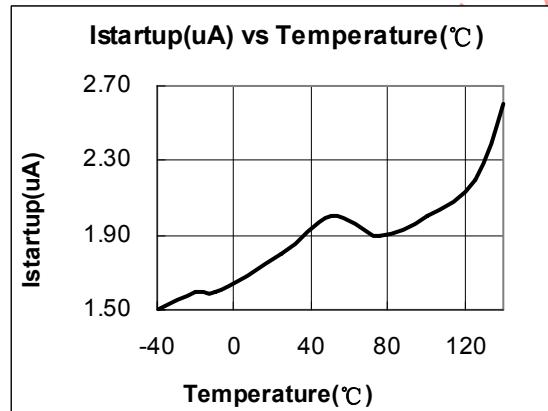
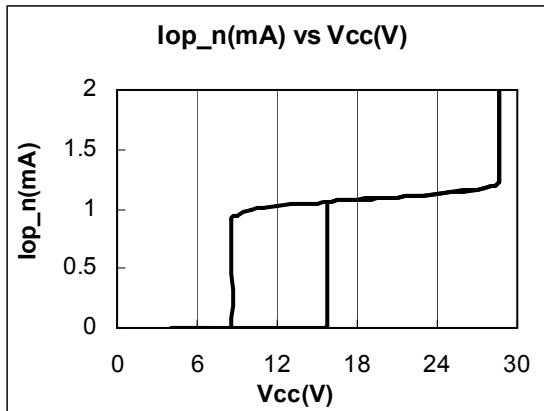
ELECTRICAL CHARACTERISTICS

($T_A = 25^\circ\text{C}$, $VCC=16\text{V}$, if not otherwise noted)

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
Supply Voltage Section (VCC Pin)						
I_VCC_Startup	VCC Start up Current	$VCC = \text{UVLO(OFF)} - 1.5\text{V}$, Measure current into VCC	-	5	15	uA
I_VCC_quiet	Operation Current without switching	FB=3V	-	2.0	4.0	mA
I_VCC_operation	Operation current with switching	FB=3V, 1nF load at GATE	-	3.0	5.0	mA
UVLO(ON)	VCC Under Voltage Lockout Enter		7.5	8.5	9.5	V
UVLO(OFF)	VCC Under Voltage Lockout Exit (Startup)		14.8	15.8	16.8	V
OVP(ON)	VCC Over Voltage Protection Enter		25	28		V
VCC_Clamp	VCC Zener Clamp Voltage	$I_{VCC} = 5\text{ mA}$		29		V
Feedback Input Section (FB Pin)						
A_{VCS}	PWM Input Gain	$\Delta V_{FB} / \Delta V_{cs}$	-	3.5	-	V/V
V _{FB_Open}	FB Open Voltage		-	5.3	-	V
I _{FB_Short}	FB pin short circuit current	Short FB pin to GND, measure current	-	1.1	-	mA
V _{TH_BM_on}	Burst Mode on threshold		-	0.9	-	V
V _{TH_BM_off}	Burst Mode off threshold		-	0.8	-	V
V _{TH_PL}	Power Limiting FB Threshold Voltage		-	4.4	-	V
T _{D_PL}	Power limiting Debounce Time		-	80	-	ms
Z _{FB_IN}	Input Impedance		-	5	-	Kohm
Current Sense Input Section (CS Pin)						
T _{blanking}	CS Input Leading Edge Blanking Time		-	300	-	ns
V _{TH_OCP_zero_ontime}	Internal current limiting threshold	@ Zero switching on time	0.42	0.45	0.48	V
V _{TH_OCP_max}	Internal current limiting threshold	@ Max. switching on time	-	0.80	-	V
V _{TH_CS_min}	Foldback mode and burst mode CS peak		-	0.3	-	V
T _{D_OC}	Over Current Detection and Control Delay	CL=1nf at GATE	-	100	-	ns
Demagnetization Detection Section (DEM Pin)						
V _{TH_DEM}	Demagnetization comparator threshold voltage		-	75	-	mV
V _{DEM_clamp(neg)}	Negative clamp voltage		-	-0.7	-	V
V _{DEM_clamp(pos)}	Positive clamp voltage		-	6.0	-	V
T _{supp}	Suppression of the transformer ringing at start of secondary stroke		-	2.5	-	us

T _{OUT}	Timeout after last demag transition		-	5	-	us
T _{DEM_delay}	Demag propagation delay		-	250	-	ns
V _{TH_OVP}	Output OVP trigger point		-	3.75	-	V
T_OVP_delay	Output OVP deglitch time constant		-	4	-	Cycle
Timer Section						
T_softstart	Internal soft startup time		-	4	-	ms
F_burst	Burst mode switching frequency		-	22	-	KHz
Fmax_QR_H	Frequency high clamp in QR mode		80	95	110	KHz
Fmin_QR_L	Frequency low clamp in QR mode		47	52	57	KHz
△F (shuffle) /F	Fmin_QR_L frequency shuffling range		-5		+5	%
Ton_max	Maximum on time		10	12.5	15	us
Toff_max	Maximum off time		50	65	85	us
G_PFM	PFM mode frequency modulation slope versus control voltage		-	90	-	KHz/V
On-chip Thermal Protection (OTP)						
T_shutdown	Thermal shutdown temperature		-	160	-	°C
Gate Drive Output (GATE Pin)						
VOL	Output Low Level	Io = 20 mA (sink)	-	-	0.8	V
VOH	Output High Level	Io = 20 mA (source)	10	-	-	V
VG_Clamp	Output Clamp Voltage Level	VCC=20V	-	16.5	-	V
T_r	Output Rising Time	CL = 1nf	-	220	-	ns
T_f	Output Falling Time	CL = 1nf	-	70	-	ns

CHARACTERIZATION PLOTS



OPERATION DESCRIPTION

Quasi-Resonant (QR) converter typically features lower EMI and higher power conversion efficiency compared to conventional hard-switched converter with a fixed switching frequency. OB2361 is a highly integrated multi-mode QR/PWM controller optimized for offline flyback converter applications. The built-in advanced energy saving with high level protection features provide cost effective solutions for energy efficient power supplies meeting EPS2.0 and European CoC.

● Startup Current and Start up Control

Startup current of OB2361 is designed to be very low so that VCC could be charged up above UVLO(OFF) threshold level and device starts up quickly. A large value startup resistor can therefore be used to minimize the power loss yet reliable startup in application. For a typical AC/DC adaptor with universal input range design, a $2\text{M}\Omega$ startup resistor could be used together with a VCC capacitor to provide a fast startup and yet low power dissipation design solution.

● Operating Current

The operating current of OB2361 is as low as 3mA. Good efficiency is achieved by the low operating current together with extended burst mode control schemes at No/light load conditions.

● Multi-Mode Operation for High Efficiency

OB2361 is a multi-mode QR/PWM controller. The controller changes the mode of operation according to line voltage and load conditions.

- At full load conditions ($\text{FB} > \text{Vth3}$, Figure 1), there are two situations: firstly, if the system input is in low line input range, the IC operates in fixed frequency CCM mode. Thus, small size transformer can be used with high power conversion efficiency. Secondly, if the system input is in high line input range, the IC operates in QR mode. In this way, high power conversion efficiency can be achieved in the universal input range when system is at full loading conditions.

- At normal operating conditions ($\text{Vth2} < \text{VFB} < \text{Vth3}$, Figure 1), the system operates in QR mode. The frequency varies depending on the line voltage and the load conditions. Therefore, the system may actually work in DCM when 95KHz frequency clamping is reached.

- At light load conditions ($\text{Vth1} < \text{VFB} < \text{Vth2}$, Figure 1), the system operates in PFM (pulse frequency modulation) mode for high power conversion

efficiency. In PFM mode, the “ON” time in a switching cycle is fixed and the system modulates the frequency according to the load conditions. Generally, in flyback converter, the decreasing of load results in voltage level decreasing at FB pin. The controller monitors the voltage level at FB and control the switching frequency. However, the valley switching characteristic is still preserved in PFM mode. That is, when load decreases, the system automatically skip more and more valleys and the switching frequency is thus reduced. In such way, a smooth frequency foldback is realized and high power conversion efficiency is achieved.

- At zero load or very light load conditions ($\text{VFB} < \text{Vth1}$), the system operates in On-Bright’s proprietary “extended burst mode”. In this condition, voltage at FB is below burst mode threshold level, Vth1 . The Gate drive output switches only when VCC voltage drops below a preset level or FB input is active to output an on state. Otherwise the gate drive remains at off state to minimize the switching loss thus reduce the standby power consumption to the greatest extend. In extended burst mode, the switching frequency is fixed to 22KHz, in this way, possible audio noise is eliminated.

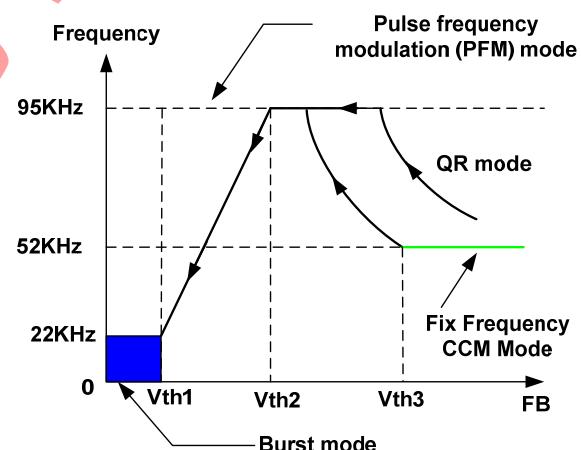


Figure 1

● Demagnetization Detection

The transformer core demagnetization is detected by monitoring the voltage activity on the auxiliary windings through DEM pin. This voltage features a flyback polarity. A new cycle starts when the power switch is activated. After the on time (determined by the CS voltage and FB), the switch is off and the flyback stroke starts. After the flyback stroke, the drain voltage shows an oscillation with a frequency of approximately

$1/2\pi\sqrt{L_p C_d}$, where L_p is the primary self inductance of primary winding of the transformer and C_d is the capacitance on the drain node.

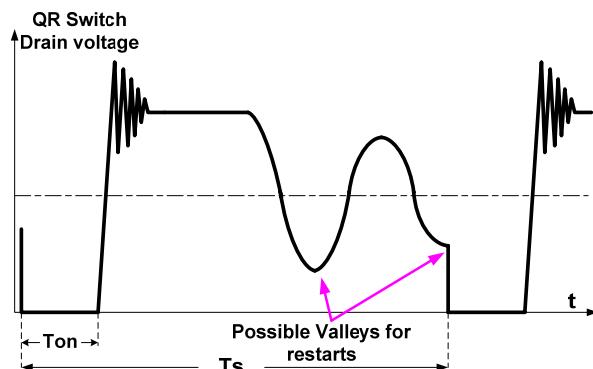


Figure 2

The typical detection level is fixed at 75mV at the DEM pin. Demagnetization is recognized by detection of a possible "valley" when the voltage at DEM is below 75mV in falling edge. DEM detection is suppressed during the ringing suppression time T_{supp} (please refer to "**Ringing Suppression Timer**" section).

- **Current Sensing and Leading Edge Blanking**

Cycle-by-Cycle current limiting is offered in OB2361. The switch current is detected by a sense resistor into the CS pin. An internal leading edge blanking circuit chops off the sense voltage spike at initial MOSFET on state, the current limit comparator is disabled and cannot turn off the external MOSFET during the blanking period.

- **Maximum and Minimum On-Time**

The minimum on-time of the system is determined by the LEB time. The IC limits the on-time to a maximum time of 12.5us.

- **Maximum Frequency Clamp**

According to the QR operation principle, the switching frequency is inversely proportional to the output power. Therefore, when the output power decreases, the switching frequency can become rather high without limiting. To meet EMI limit and to achieve high efficiency at light loading conditions, the maximum switching frequency in OB2361 is internally limited to 95KHz.

- **Ringing Suppression Timer**

A ringing suppression timer T_{supp} is implemented in OB2361. In normal operation, T_{supp} starts when CS reaches the feedback voltage FB, the

external power switch is set to off state. During T_{supp} , the external power switch remains in off state and cannot be turned on again. The ringing suppression time is necessary in applications where the transformer has a large leakage inductance. In OB2361, the ringing suppression timer T_{supp} is set to 2.5us internally.

- **OCP Compensation**

The variation of max output power in QR system can be rather large if no compensation is provided. The OCP threshold value is self adjusted lower at higher AC voltage. This OCP threshold slope adjustment helps to compensate the increased output power limit at higher AC voltage. In OB2361, a proprietary OCP compensation block is integrated and no external components are needed. The OCP threshold in OB2361 is a function of the switching ON time. For the ON time less than 12.5us, the OCP threshold changes linearly from 0.45V to 0.8V. For the ON time larger than 12.5us, the OCP threshold is clamped to 0.8V, as shown in Figure 3.

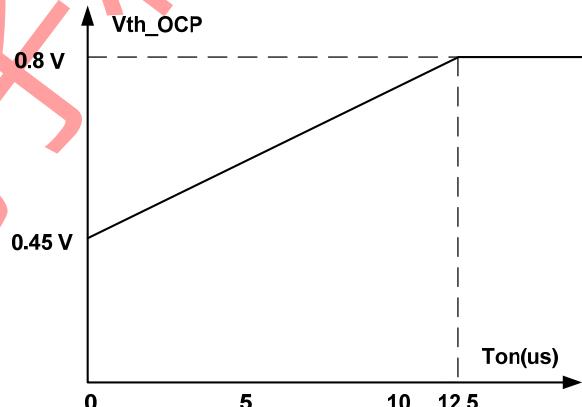


Figure 3

- **Adaptive Frequency Shuffling and Slope Compensation**

In OB2361, when the fixed frequency CCM mode is reached, the frequency shuffling and slope compensation will be automatically added to the system to improve the EMI performance and current loop stability. The magnitude of shuffling lies in the range of $\pm 5\%$ of 52K Hz. When the system exits fixed frequency CCM mode, the frequency shuffling and slope compensation will automatically disappear.

- **Output Over Voltage Protection (OVP)**

An output over voltage protection (OVP) is implemented by sensing the auxiliary winding voltage at DEM pin during the flyback phase. The auxiliary winding voltage is a well-defined replica of the output voltage. The OVP works by sampling

the plateau voltage at DEM pin during the flyback phase. Internal 2.5us delay (plateau sensing) guarantees a clean plateau, provided that the leakage inductance ringing has been fully damped.

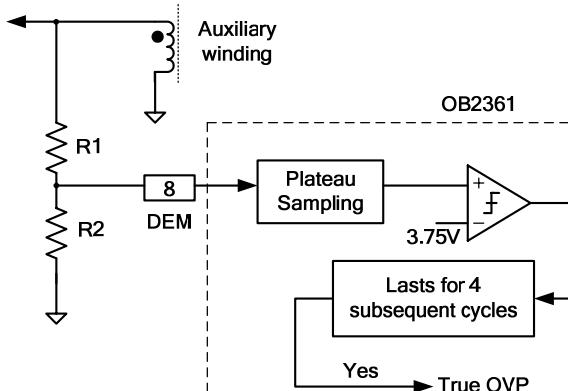


Figure 4

If the sampled plateau voltage exceeds the OVP trip level (3.75V), an internal counter starts counting subsequent OVP events. If OVP events are detected in successive 4 cycles, the controller assumes a true OVP and stops all switching operations. The counter has been added to prevent incorrect OVP detection which might occur during ESD or lightning events. If the output voltage exceeds the OVP trip level less than 4

successive cycles, the internal counter will be cleared and no fault is asserted.

● Over Load Protection

When over load (for example, short circuit) occurs, the feedback current of photo coupler is below minimum value, FB pin voltage approaches its floating pin voltage, then a fault is detected. If this fault is present for more than 80ms, the controller enters an auto-recovery soft burst mode. All pulses are stopped, VCC will drops below UVLO and the controller will try to restart with the power on soft start. It resumes operation once the fault disappears.

● Pin Floating and Short Protection

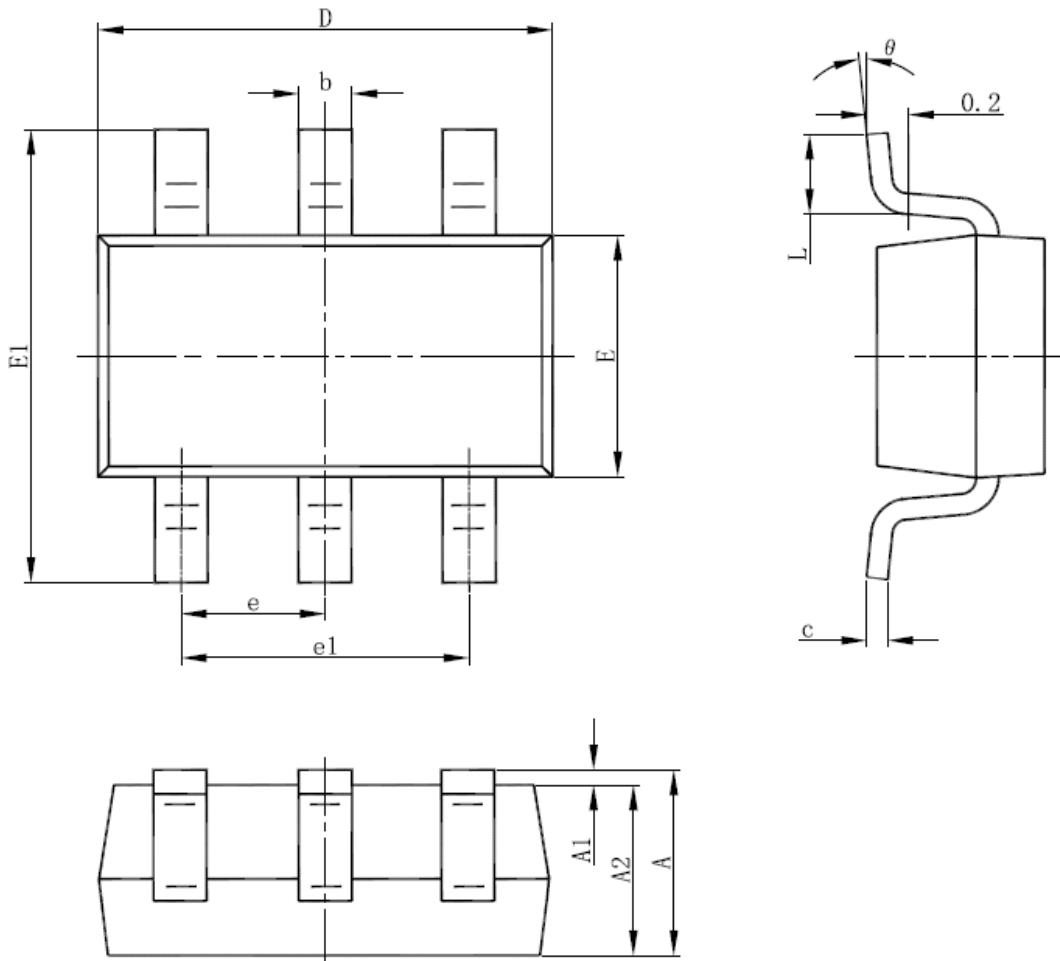
OB2361 provides pin floating protection for CS, FB, DEM, etc., In cases when the pins are floating, Gate switching is disabled, thus protect the power system.

● Gate Drive

The Gate pin is connected to the gate of an external power switch. An internal 16.5V clamp is added for MOSFET gate protection at high VCC voltage. When VCC voltage drops below UVLO(ON), the Gate pin is internally pull low to maintain the off state.

PACKAGE MECHANICAL DATA

SOT-23-6L PACKAGE OUTLINE DIMENSIONS



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.000	1.300	0.039	0.051
A1	0.000	0.150	0.000	0.006
A2	1.000	1.200	0.039	0.047
b	0.300	0.500	0.012	0.020
c	0.100	0.200	0.004	0.008
D	2.800	3.020	0.110	0.119
E	1.500	1.700	0.059	0.067
E1	2.600	3.000	0.102	0.118
e	0.950 (BSC)		0.037 (BSC)	
e1	1.800	2.000	0.071	0.079
L	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°

IMPORTANT NOTICE

RIGHT TO MAKE CHANGES

On-Bright Electronics Corp. reserves the right to make corrections, modifications, enhancements, improvements and other changes to its products and services at any time and to discontinue any product or service without notice. Customers should obtain the latest relevant information before placing orders and should verify that such information is current and complete.

WARRANTY INFORMATION

On-Bright Electronics Corp. warrants performance of its hardware products to the specifications applicable at the time of sale in accordance with its standard warranty. Testing and other quality control techniques are used to the extent it deems necessary to support this warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily performed. On-Bright Electronics Corp. assumes no liability for application assistance or customer product design. Customers are responsible for their products and applications using On-Bright's components, data sheet and application notes. To minimize the risks associated with customer products and applications, customers should provide adequate design and operating safeguards.

LIFE SUPPORT

On-Bright Electronics Corp.'s products are not designed to be used as components in devices intended to support or sustain human life. On-Bright Electronics Corp. will not be held liable for any damages or claims resulting from the use of its products in medical applications.

MILITARY

On-Bright Electronics Corp.'s products are not designed for use in military applications. On-Bright Electronics Corp. will not be held liable for any damages or claims resulting from the use of its products in military applications.